

# Chapter 6 – Arithmetic

6.1. Overflow cases are specifically indicated. In all other cases, no overflow occurs.

$\begin{array}{r} 010110 \\ + 001001 \\ \hline 011111 \end{array}$	$\begin{array}{r} (+22) \\ + (+9) \\ \hline (+31) \end{array}$	$\begin{array}{r} 101011 \\ + 100101 \\ \hline 010000 \\ \text{overflow} \end{array}$	$\begin{array}{r} (-21) \\ + (-27) \\ \hline (-48) \end{array}$	$\begin{array}{r} 111111 \\ + 000111 \\ \hline 000110 \end{array}$	$\begin{array}{r} (-1) \\ + (+7) \\ \hline (+6) \end{array}$
$\begin{array}{r} 011001 \\ + 010000 \\ \hline 101001 \\ \text{overflow} \end{array}$	$\begin{array}{r} (+25) \\ + (+16) \\ \hline (+41) \end{array}$	$\begin{array}{r} 110111 \\ + 111001 \\ \hline 110000 \end{array}$	$\begin{array}{r} (-9) \\ + (-7) \\ \hline (-16) \end{array}$	$\begin{array}{r} 010101 \\ + 101011 \\ \hline 000000 \end{array}$	$\begin{array}{r} (+21) \\ + (-21) \\ \hline (0) \end{array}$
$\begin{array}{r} 010110 \\ - 011111 \\ \hline \end{array}$	$\begin{array}{r} (+22) \\ - (+31) \\ \hline (-9) \end{array}$	$\begin{array}{r} 010110 \\ + 100001 \\ \hline 110111 \end{array}$			
$\begin{array}{r} 111110 \\ - 100101 \\ \hline \end{array}$	$\begin{array}{r} (-2) \\ - (-27) \\ \hline (+25) \end{array}$	$\begin{array}{r} 111110 \\ + 011011 \\ \hline 011001 \end{array}$			
$\begin{array}{r} 100001 \\ - 011101 \\ \hline \end{array}$	$\begin{array}{r} (-31) \\ - (+29) \\ \hline (-60) \end{array}$	$\begin{array}{r} 100001 \\ + 100011 \\ \hline 000100 \\ \text{overflow} \end{array}$			
$\begin{array}{r} 111111 \\ - 000111 \\ \hline \end{array}$	$\begin{array}{r} (-1) \\ - (+7) \\ \hline (-8) \end{array}$	$\begin{array}{r} 111111 \\ + 111001 \\ \hline 111000 \end{array}$			
$\begin{array}{r} 000111 \\ - 111000 \\ \hline \end{array}$	$\begin{array}{r} (+7) \\ - (-8) \\ \hline (+15) \end{array}$	$\begin{array}{r} 000111 \\ + 001000 \\ \hline 001111 \end{array}$			
$\begin{array}{r} 011010 \\ - 100010 \\ \hline \end{array}$	$\begin{array}{r} (+26) \\ - (-30) \\ \hline (+56) \end{array}$	$\begin{array}{r} 011010 \\ + 011110 \\ \hline 111000 \\ \text{overflow} \end{array}$			

6.2. (a) In the following answers, rounding has been used as the truncation method (see Section 6.7.3) when the answer cannot be represented exactly in the signed 6-bit format.

0.5:	010000	all cases
−0.123:	100100	Sign-and-magnitude
	111011	1's-complement
	111100	2's-complement
−0.75:	111000	Sign-and-magnitude
	100111	1's-complement
	101000	2's-complement
−0.1:	100011	Sign-and-magnitude
	111100	1's-complement
	111101	2's-complement

(b)

$e = 2^{-6}$  (assuming rounding, as in (a))

$e = 2^{-5}$  (assuming chopping or Von Neumann rounding)

(c) assuming rounding:

(a)	3
(b)	6
(c)	9
(d)	19

6.3. The two ternary representations are given as follows:

Sign-and-magnitude	3's-complement
+11011	011011
−10222	212001
+2120	002120
−1212	221011
+10	000010
−201	222022

6.4. Ternary numbers with addition and subtraction operations:

Decimal Sign-and-magnitude	Ternary Sign-and-magnitude	Ternary 3's-complement
56	+2002	002002
-37	-1101	221122
122	11112	011112
-123	-11120	211110

Addition operations:

002002	002002	002002
+ 221122	+ 011112	+ 211110
<hr/> 000201	<hr/> 020121	<hr/> 220112
221122	221122	011112
+ 011112	+ 211110	+ 211110
<hr/> 010011	<hr/> 210002	<hr/> 222222

Subtraction operations:

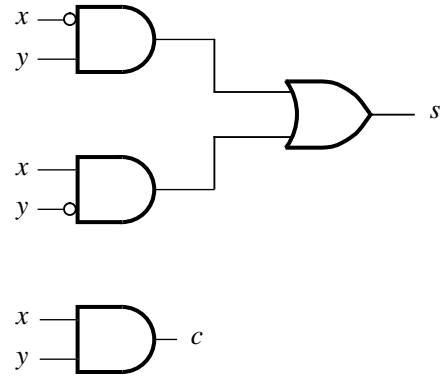
002002	002002
- 221122	+ 001101
<hr/> 010110	
002002	002002
- 011112	+ 211111
<hr/> 220120	
002002	002002
- 211110	+ 011120
<hr/> 020122	
221122	221122
- 011112	+ 211111
<hr/> 210010	
221122	221122
- 211110	+ 011120
<hr/> 010012	
011112	011112
- 211110	+ 011120
<hr/> 100002	
	overflow

6.5. (a)

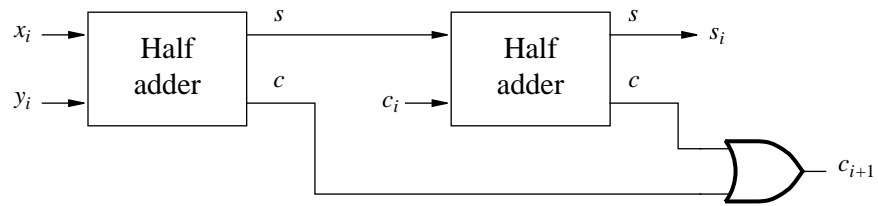
$x$	$y$	$s$	$c$
0	0	0	0
0	1	1	0
1	0	1	0
1	1	0	1

$$s = x \oplus y$$

$$c = xy$$



(b)



(c) The longest path through the circuit in Part (b) is 6 gate delays (including input inversions) in producing  $s_i$ ; and the longest path through the circuit in Figure 6.2a is 3 gate delays in producing  $s_i$ , assuming that  $s_i$  is implemented as a two-level AND-OR circuit, and including input inversions.

- 6.6. Assume that the binary integer is in memory location `BINARY`, and the string of bytes representing the answer starts at memory location `DECIMAL`, high-order digits first.

**68000 Program:**

	<code>MOVE</code>	<code>#10,D2</code>	
	<code>CLR.L</code>	<code>D1</code>	
	<code>MOVE</code>	<code>BINARY,D1</code>	Get binary number; note that high-order word in D1 is still zero.
	<code>MOVE.B</code>	<code>#4,D3</code>	Use D3 as counter.
<code>LOOP</code>	<code>DIVU</code>	<code>D2,D1</code>	Leaves quotient in low half of D1 and remainder in high half of D1.
	<code>SWAP</code>	<code>D1</code>	
	<code>MOVE.B</code>	<code>D1,DECIMAL(D3)</code>	
	<code>CLR</code>	<code>D1</code>	Clears low half of D1.
	<code>SWAP</code>	<code>D1</code>	
	<code>DBRA</code>	<code>D3,LOOP</code>	

**IA-32 Program:**

	<code>MOV</code>	<code>EBX,10</code>	
	<code>MOV</code>	<code>EAX,BINARY</code>	Get binary number.
	<code>LEA</code>	<code>EDI,DECIMAL</code>	
	<code>DEC</code>	<code>EDI</code>	
	<code>MOV</code>	<code>ECX,5</code>	Load counter ECX.
<code>LOOPSTART:</code>	<code>DIV</code>	<code>EBX</code>	<code>[EAX]/[EBX]</code> ; quotient in EAX and remainder in EDX.
	<code>MOV</code>	<code>[EDI + ECX],DL</code>	
	<code>LOOP</code>	<code>LOOPSTART</code>	

6.7. The ARM and IA-32 subroutines both use the following algorithm to convert the four-digit decimal integer  $D_3D_2D_1D_0$  (each  $D_i$  is in BCD code) into binary:

- Move  $D_0$  into register REG.
- Multiply  $D_1$  by 10.
- Add product into REG.
- Multiply  $D_2$  by 100.
- Add product into REG.
- Multiply  $D_3$  by 1000.
- Add product into REG.

(i) The ARM subroutine assumes that the addresses DECIMAL and BINARY are passed to it on the processor stack in positions param1 and param2 as shown in Figure 3.13. The subroutine first saves registers and sets up the frame pointer FP (R12).

**ARM Subroutine:**

CONVERT	STMFD	SP!,{R0–R6,FP,LR}	Save registers.
	ADD	FP,SP,#28	Load frame pointer.
	LDR	R0,[FP,#8]	Load R0 and R1
	LDR	R0,[R0]	with decimal digits.
	MOV	R1,R0	
	AND	R0,R0,#&F	[R0] = $D_0$ .
	MOV	R2,#&F	Load mask bits into R2.
	MOV	R4,#10	Load multipliers
	MOV	R5,#100	into R4, R5, and R6.
	MOV	R6,#1000	
	AND	R3,R2,R1,LSR #4	Get $D_1$ into R3.
	MLA	R0,R3,R4,R0	Add $10D_1$ into R0.
	AND	R3,R2,R1,LSR #8	Get $D_2$ into R3.
	MLA	R0,R3,R5,R0	Add $100D_2$ into R0.
	AND	R3,R2,R1,LSR #12	Get $D_3$ into R3.
	MLA	R0,R3,R6,R0	Add $1000D_3$ into R0.
	LDR	R1,[FP,#12]	Store converted value
	STR	R0,[R1]	into BINARY.
	LDMFD	SP!,{R0–R6,FP,PC}	Restore registers
			and return.

(ii) The IA-32 subroutine assumes that the addresses DECIMAL and BINARY are passed to it on the processor stack in positions param1 and param2 as shown in Figure 3.48. The subroutine first sets up the frame pointer EBP, and then allocates and initializes the local variables 10, 100, and 1000, on the stack.

#### IA-32 Subroutine:

CONVERT:	PUSH	EBP	Set up frame
	MOV	EBP,ESP	pointer.
	PUSH	10	Allocate and initialize
	PUSH	100	local variables.
	PUSH	1000	
	PUSH	EDX	Save registers.
	PUSH	ESI	
	PUSH	EAX	
	MOV	EDX,[EBP + 8]	Load four decimal
	MOV	EDX,[EDX]	digits into
	MOV	ESI,EDX	EDX and ESI.
	AND	EDX,FH	[EDX] = $D_0$ .
	SHR	ESI,4	
	MOV	EAX,ESI	
	AND	EAX,FH	
	MUL	[EBP - 4]	
	ADD	EDX,EAX	[EDX] = binary of $D_1D_0$ .
	SHR	ESI,4	
	MOV	EAX,ESI	
	AND	EAX,FH	
	MUL	[EBP - 8]	
	ADD	EDX,EAX	[EDX] = binary of $D_2D_1D_0$ .
	SHR	ESI,4	
	MOV	EAX,ESI	
	AND	EAX,FH	
	MUL	[EBP - 12]	
	ADD	EDX,EAX	[EDX] = binary of $D_3D_2D_1D_0$ .
	MOV	EAX,[EBP + 12]	Store converted
	MOV	[EAX],EDX	value into BINARY.
	POP	EAX	Restore registers.
	POP	ESI	
	POP	EDX	
	ADD	ESP,12	Remove local parameters.
	POP	EBP	Restore EBP.
	RET		Return.

(iii) The 68000 subroutine uses a loop structure to convert the four-digit decimal integer  $D_3D_2D_1D_0$  (each  $D_i$  is in BCD code) into binary. At the end of successive passes through the loop, register D0 contains the accumulating values  $D_3$ ,  $10D_3 + D_2$ ,  $100D_3 + 10D_2 + D_1$ , and  $\text{binary} = 1000D_3 + 100D_2 + 10D_1 + D_0$ . Assume that DECIMAL is the address of a 16-bit word containing the four BCD digits, and that BINARY is the address of a 16-bit word that is to contain the converted binary value.

The addresses DECIMAL and BINARY are passed to the subroutine in registers A0 and A1.

**68000 Subroutine:**

CONVERT	MOVEM.L	D0–D2,–(A7)	Save registers.
	CLR.L	D0	
	CLR.L	D1	
	MOVE.W	(A0),D1	Load four decimal digits into D1.
LOOP	MOVE.B	#3,D2	Load counter D3.
	MULU.W	#10,D0	Multiply accumulated value in D0 by 10.
	ASL.L	#4,D1	Bring next $D_i$ digit into low half of D1.
	SWAP.W	D1	into low half of D1.
	ADD.W	D1,D0	Add into accumulated value in D0.
	CLR.W	D1	Clear out current digit and bring remaining digits into low half of D1.
	SWAP.W	D1	
	DBRA	D2,LOOP	Check if done.
	MOVE.W	D0,(A1)	Store binary result in BINARY.
	MOVEM.L	(A7)+,D0–D2	Restore registers.
	RTS		Return.



6.8. (a) The output carry is 1 when  $A + B \geq 10$ . This is the condition that requires the further addition of  $6_{10}$ .

(b)

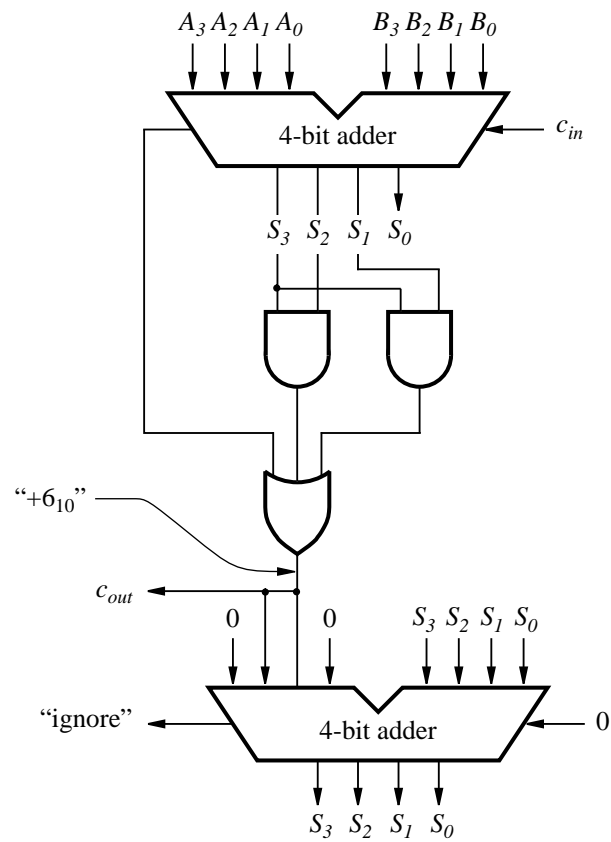
$$(1) \quad \begin{array}{r} 0101 \\ + 0110 \\ \hline 1011 \end{array} > 10_{10} \quad \begin{array}{r} 5 \\ + 6 \\ \hline 11 \end{array}$$

$$\begin{array}{r} + 0110 \\ \hline 0001 \end{array}$$

output carry = 1

$$(2) \quad \begin{array}{r} 0011 \\ + 0100 \\ \hline 0111 \end{array} < 10_{10} \quad \begin{array}{r} 3 \\ + 4 \\ \hline 7 \end{array}$$

(c)



6.9. Consider the truth table in Figure 6.1 for the case  $i = n - 1$ , that is, for the sign bit position. Overflow occurs only when  $x_{n-1}$  and  $y_{n-1}$  are the same and  $s_{n-1}$  is different. This occurs in the second and seventh rows of the table; and  $c_n$  and  $c_{n-1}$  are different only in those rows. Therefore,  $c_n \oplus c_{n-1}$  is a correct indicator of overflow.

6.10. (a) The additional logic is defined by the logic expressions:

$$\begin{aligned}
c_{16} &= G_0^{II} + P_0^{II} c_0 \\
c_{32} &= G_1^{II} + P_1^{II} G_0^{II} + P_1^{II} P_0^{II} c_0 \\
c_{48} &= G_2^{II} + P_2^{II} G_1^{II} + P_2^{II} P_1^{II} G_0^{II} + P_2^{II} P_1^{II} P_0^{II} c_0 \\
c_{64} &= G_3^{II} + P_3^{II} G_2^{II} + P_3^{II} P_2^{II} G_1^{II} + P_3^{II} P_2^{II} P_1^{II} G_0^{II} + P_3^{II} P_2^{II} P_1^{II} P_0^{II} c_0
\end{aligned}$$

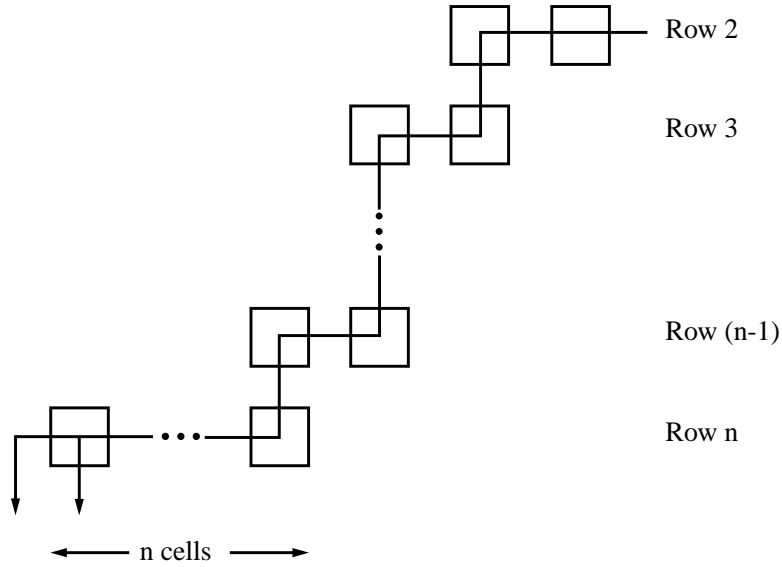
This additional logic is identical in form to the logic inside the lookahead circuit in Figure 6.5. (Note that the outputs  $c_{16}$ ,  $c_{32}$ ,  $c_{48}$ , and  $c_{64}$ , produced by the 16-bit adders are not needed because those outputs are produced by the additional logic.)

(b) The inputs  $G_i^{II}$  and  $P_i^{II}$  to the additional logic are produced after 5 gate delays, the same as the delay for  $c_{16}$  in Figure 6.5. Then all outputs from the additional logic, including  $c_{64}$ , are produced 2 gate delays later, for a total of 7 gate delays. The carry input  $c_{48}$  to the last 16-bit adder is produced after 7 gate delays. Then  $c_{60}$  into the last 4-bit adder is produced after 2 more gate delays, and  $c_{63}$  is produced after another 2 gate delays inside that 4-bit adder. Finally, after one more gate delay (an XOR gate),  $s_{63}$  is produced with a total of  $7 + 2 + 2 + 1 = 12$  gate delays.

(c) The variables  $s_{31}$  and  $c_{32}$  are produced after 12 and 7 gate delays, respectively, in the 64-bit adder. These two variables are produced after 10 and 7 gate delays in the 32-bit adder, as shown in Section 6.2.1.

- 6.11. (a) Each B cell requires 3 gates as shown in Figure 6.4a. The carries  $c_1$ ,  $c_2$ ,  $c_3$ , and  $c_4$ , require 2, 3, 4, and 5, gates, respectively; and the outputs  $G_0^I$  and  $P_0^I$  require 4 and 1 gates, as seen from the logic expressions in Section 6.2.1. Therefore, a total of  $12 + 19 = 31$  gates are required for the 4-bit adder.
- (b) Four 4-bit adders require  $4 \times 31 = 124$  gates, and the carry-lookahead logic block requires 19 gates because it has the same structure as the lookahead block in Figure 6.4. Total gate count is thus 143. However, we should subtract  $4 \times 5 = 20$  gates from this total corresponding to the logic for  $c_4$ ,  $c_8$ ,  $c_{12}$ , and  $c_{16}$ , that is in the 4-bit adders but which is replaced by the lookahead logic in Figure 6.5. Therefore, total gate count for the 16-bit adder is  $143 - 20 = 123$  gates.

6.12. The worst case delay path is shown in the following figure:



Each of the two FA blocks in rows 2 through  $n - 1$  introduces 2 gate delays, for a total of  $4(n - 2)$  gate delays. Row  $n$  introduces  $2n$  gate delays. Adding in the initial AND gate delay for row 1 and all other cells, total delay is:

$$4(n - 2) + 2n + 1 = 6n - 8 + 1 = 6(n - 1) - 1$$

6.13. The solutions, including decimal equivalent checks, are:

$$\begin{array}{rcl}
 \text{B} & = & 00101 \quad (5) \\
 \times \text{A} & = & \begin{array}{r} 10101 \\ 00101 \\ 0 \\ 00101 \\ 001010 \\ \hline 001101001 \end{array} \quad \begin{array}{r} \times (21) \\ \hline (105) \end{array} \\
 & & (105)
 \end{array}$$

$$\begin{array}{rcl}
 & & 100 \\
 00101 & \sqrt{10101} & 5 \sqrt{21} \\
 & \begin{array}{r} 101 \\ \hline 00001 \end{array} & \begin{array}{r} 4 \\ 20 \\ \hline 1 \end{array}
 \end{array}$$

6.14. The multiplication and division charts are:

$A \times B :$

	M		
	00101		
0	00000	10101	Initial configuration
C	A	Q	
0	00101	10101	1st cycle
0	00010	11010	
0	00010	11010	2nd cycle
0	00001	01101	
0	00110	01101	3rd cycle
0	00011	00110	
0	00011	00110	4th cycle
0	00001	10011	
0	00110	10011	5th cycle
0	00011	01001	
	product		

$A / B :$

	000000	10101	
	A	Q	
	000101		Initial configuration
	M		
shift	000001	0 1 0 1	
subtract	111011		1st cycle
	111100	0 1 0 1	
shift	111000	1 0 1	
add	000101		2nd cycle
	111101	1 0 1	
shift	111011	0 1	
add	000101		3rd cycle
	000000	0 1	
shift	000000	1	
subtract	111011		4th cycle
	111011	1	
shift	110111	0 0 1	
add	000101		5th cycle
	111100	0 0 1	
add	000101		
	000001	quotient	
	remainder		

### 6.15. ARM Program:

Use R0 as the loop counter.

	MOV	R1,#0	
	MOV	R0,#32	
LOOP	TST	R2,#1	Test LSB of multiplier.
	ADDNE	R1,R3,R1	Add multiplicand if LSB = 1.
	MOV	R1,R1,RRX	Shift [R1] and [R2] right
	MOV	R2,R2,RRX	one bit position, with [C].
	SUBS	R0,R0,#1	Check if done.
	BGT	LOOP	

### 68000 program:

Assume that D2 and D3 contain the multiplier and the multiplicand, respectively. The high- and low-order halves of the product will be stored in D1 and D2. Use D0 as the loop counter.

	CLR.L	D1	
	MOVE.B	#31,D0	
LOOP	ANDI.W	#1,D2	Test LSB of multiplier.
	BEQ	NOADD	
	ADD.L	D3,D1	Add multiplicand if LSB = 1.
NOADD	ROXR.L	#1,D1	Shift [D1] and [D2] right
	ROXR.L	#1,D2	one bit position, with [C].
	DBRA	D0,LOOP	Check if done.

### IA-32 Program:

Use registers EAX, EDX, and EDI, as  $R_1$ ,  $R_2$ , and  $R_3$ , respectively, and use ECX as the loop counter.

	MOV	EAX,0	
	MOV	ECX,32	
	SHR	EDX,1	Set [CF] = LSB of multiplier.
LOOPSTART:	JNC	NOADD	
	ADD	EAX,EDI	Add multiplicand if LSB = 1.
NOADD:	RCR	EAX,1	Shift [EAX] and [EDX] right
	RCR	EDX,1	one bit position, with [CF].
	LOOP	LOOPSTART	Check if done.

#### 6.16. ARM Program:

Use the register assignment R1, R2, and R0, for the dividend, divisor, and remainder, respectively. As computation proceeds, the quotient will be shifted into R1.

	MOV	R0,#0	Clear R0.
	MOV	R3,#32	Initialize counter R3.
LOOP	MOVS	R1,R1,LSL #1	Two-register left
	ADCS	R0,R0,R0	shift of R0 and R1
			by one position.
	SUBCCS	R0,R0,R2	Implement step 1
	ADDCSS	R0,R0,R2	of the algorithm.
	ORRPL	R1,R1,#1	
	SUBS	R3,R3,#1	Check if done.
	BGT	LOOP	
	TST	R0,R0	Implement step 2
	ADDMI	R0,R2,R0	of the algorithm.

#### 68000 Program:

Assume that D1 and D2 contain the dividend and the divisor, respectively. We will use D0 to store the remainder. As computation proceeds, the quotient will be shifted into D1.

	CLR	D0	Clear D0.
	MOVE.B	#15,D3	Initialize counter D3.
LOOP	ASL	#1,D1	Two-register left shift of
	ROXL	#1,D0	D0 and D1 by one position.
	BCS	NEGRM	Implement step 1
	SUB	D2,D0	of the algorithm.
	BRA	SETQ	
NEGRM	ADD	D2,D0	
SETQ	BMI	COUNT	
	ORI	#1,D1	
COUNT	DBRA	D3,LOOP	Check if done.
	TST	D0	Implement step 2
	BPL	DONE	of the algorithm.
	ADD	D2,D0	
DONE	...		

**IA-32 Program:**

Use the register assignment EAX, EBX, and EDX, for the dividend, divisor, and remainder, respectively. As computation proceeds, the quotient is shifted into EAX.

	MOV	EDX,0	Clear EDX.
	MOV	ECX,32	Initialize counter ECX.
LOOPSTART:	SHL	EAX,1	Two-register left
	RCL	EDX,1	shift of EDX and EAX
			by one position.
	JC	NEGRM	Implement step 1
	SUB	EDX,EBX	of the algorithm.
	JMP	SETQ	
NEGRM:	ADD	EDX,EBX	
SETQ:	JS	COUNT	
	OR	EAX,1	
COUNT:	LOOP	LOOPSTART	Check if done.
	TEST	EDX,EDX	Implement step 2
	JNS	DONE	of the algorithm.
	ADD	EDX,EBX	
DONE:	...		



sign  
extension

sign  
extension

sign  
extension

6.18. The multiplication answers are:

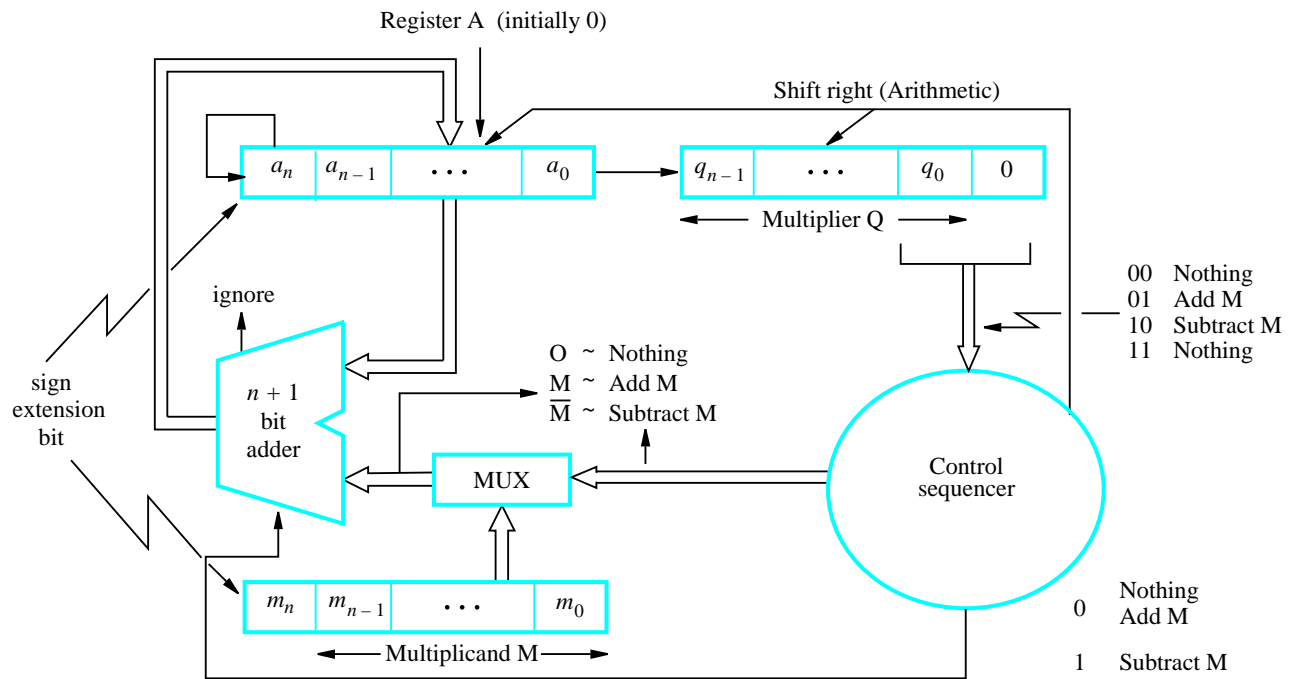
$$\begin{array}{r}
 (a) \quad \begin{array}{r} 010111 \\ \times 110110 \\ \hline \end{array} \qquad \begin{array}{r} \phantom{00000000} 0 \ 1 \ 0 \ 1 \ 1 \ 1 \\ \phantom{00000000} -1 \ +2 \ -2 \\ \hline 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 0 \ 1 \ 0 \ 0 \ 1 \ 0 \\ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 1 \ 1 \ 1 \ 0 \\ \hline 1 \ 1 \ 1 \ 1 \ 1 \ 0 \ 2 \ 1 \ 1 \ 0 \ 1 \ 1 \\ \hline 1 \ 1 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0 \end{array}
 \end{array}$$

$$\begin{array}{r}
 (b) \quad \begin{array}{r} 110011 \\ \times 101100 \\ \hline \end{array} \qquad \begin{array}{r} \phantom{00000000} 1 \ 1 \ 0 \ 0 \ 1 \ 1 \\ \phantom{00000000} -1 \ -1 \ 0 \\ \hline 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \\ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 0 \ 1 \ 1 \\ \hline 0 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 0 \end{array}
 \end{array}$$

$$\begin{array}{r}
 (c) \quad \begin{array}{r} 110101 \\ \times 011011 \\ \hline \end{array} \qquad \begin{array}{r} \phantom{00000000} 1 \ 1 \ 0 \ 1 \ 0 \ 1 \\ \phantom{00000000} +2 \ -1 \ -1 \\ \hline 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 1 \ 1 \\ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 1 \ 1 \\ \hline 1 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 1 \ 1 \\ \hline 1 \ 1 \ 1 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 1 \ 1 \end{array}
 \end{array}$$

$$\begin{array}{r}
 (d) \quad \begin{array}{r} 001111 \\ \times 001111 \\ \hline \end{array} \qquad \begin{array}{r} \phantom{00000000} 0 \ 0 \ 1 \ 1 \ 1 \ 1 \\ \phantom{00000000} +1 \ -1 \\ \hline 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1 \\ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 1 \\ \hline 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1 \end{array}
 \end{array}$$

- 6.19. Both the A and M registers are augmented by one bit to the left to hold a sign extension bit. The adder is changed to an  $n + 1$ -bit adder. A bit is added to the right end of the Q register to implement the Booth multiplier recoding operation. It is initially set to zero. The control logic decodes the two bits at the right end of the Q register according to the Booth algorithm, as shown in the following logic circuit. The right shift is an arithmetic right shift as indicated by the repetition of the extended sign bit at the left end of the A register. (The only case that actually requires the sign extension bit is when the  $n$ -bit multiplicand is the value  $-2^{(n-1)}$ ; for all other operands, the A and M registers could have been  $n$ -bit registers and the adder could have been an  $n$ -bit adder.)



6.20 (a)

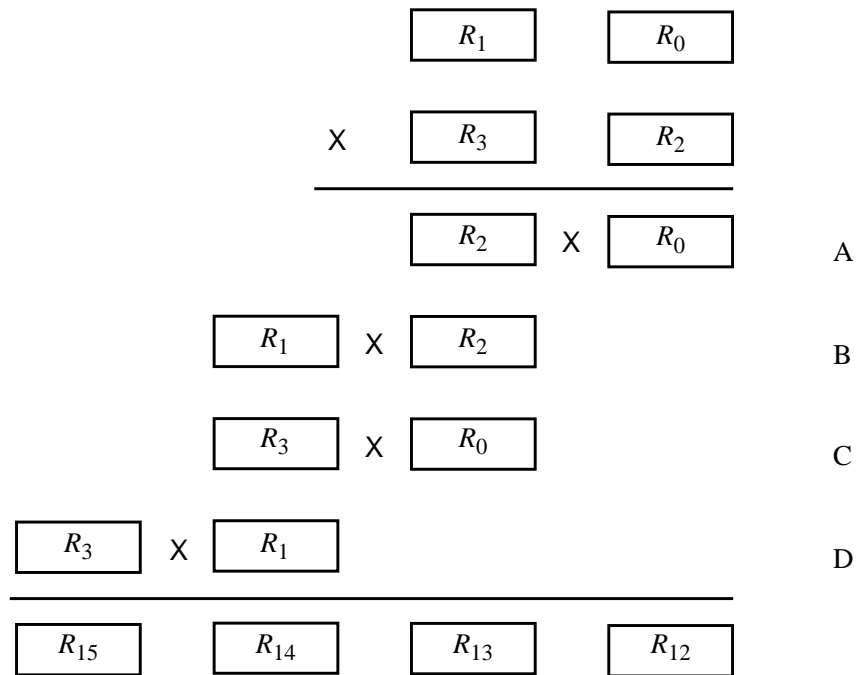
$$\begin{array}{r}
 1110 \\
 \times 1101 \\
 \hline
 1110 \\
 0000 \\
 1000 \\
 0000 \\
 \hline
 0110
 \end{array}
 \qquad
 \begin{array}{r}
 -2 \\
 \times -3 \\
 \hline
 6
 \end{array}$$

(b)

$$\begin{array}{r}
 0010 \\
 \times 1110 \\
 \hline
 0000 \\
 0100 \\
 1000 \\
 0000 \\
 \hline
 1100
 \end{array}
 \qquad
 \begin{array}{r}
 2 \\
 \times -2 \\
 \hline
 -4
 \end{array}$$

This technique works correctly for the same reason that modular addition can be used to implement signed-number addition in the 2's-complement representation, because multiplication can be interpreted as a sequence of additions of the multiplicand to shifted versions of itself.

- 6.21. The four 32-bit subproducts needed to generate the 64-bit product are labeled A, B, C, and D, and shown in their proper shifted positions in the following figure:



The 64-bit product is the sum of A, B, C, and D. Using register transfers and multiplication and addition operations executed by the arithmetic unit described, the 64-bit product is generated without using any extra registers by the following steps:

$$\begin{aligned}
 R_{12} &\leftarrow [R_0] \\
 R_{13} &\leftarrow [R_2] \\
 R_{14} &\leftarrow [R_1] \\
 R_{15} &\leftarrow [R_3] \\
 R_3 &\leftarrow [R_{14}] \\
 R_1 &\leftarrow [R_{15}] \\
 R_{13}, R_{12} &\leftarrow [R_{13}] \times [R_{12}] \\
 R_{15}, R_{14} &\leftarrow [R_{15}] \times [R_{14}] \\
 R_3, R_2 &\leftarrow [R_3] \times [R_2] \\
 R_1, R_0 &\leftarrow [R_1] \times [R_0] \\
 R_{13} &\leftarrow [R_2] \text{ Add } [R_{13}] \\
 R_{14} &\leftarrow [R_3] \text{ Add with carry } [R_{14}] \\
 R_{15} &\leftarrow 0 \text{ Add with carry } [R_{15}] \\
 R_{13} &\leftarrow [R_0] \text{ Add } [R_{13}] \\
 R_{14} &\leftarrow [R_1] \text{ Add with carry } [R_{14}] \\
 R_{15} &\leftarrow 0 \text{ Add with carry } [R_{15}]
 \end{aligned}$$

This procedure destroys the original contents of the operand registers. Steps 5 and 6 result in swapping the contents of  $R_1$  and  $R_3$  so that subproducts B and C can be computed in adjacent register pairs. Steps 11, 12, and 13, add the subproduct B into the 64-bit product registers; and steps 14, 15, and 16, add the subproduct C into these registers.

- 6.22. (a) The worst case delay path in Figure 6.16a is along the staircase pattern that includes the two FA blocks at the right end of each of the first two rows (a total of four FA block delays), followed by the four FA blocks in the third row. Total delay is therefore 17 gate delays, including the initial AND gate delay to develop all bit products.

In Figure 6.16b, the worst case delay path is vertically through the first two rows (a total of two FA block delays), followed by the four FA blocks in the third row for a total of 13 gate delays, including the initial AND gate delay to develop all bit products.

(b) Both arrays are  $4 \times 4$  cases.

Note that 17 is the result of applying the expression  $6(n - 1) - 1$  with  $n = 4$  for the array in Figure 6.16a.

A similar expression for the Figure 6.16b array is developed as follows. The delay through  $(n - 2)$  carry-save rows of FA blocks is  $2(n - 2)$  gate delays, followed by  $2n$  gate delays along the  $n$  FA blocks of the last row, for a total of

$$2(n - 2) + 2n + 1 = 4(n - 1) + 1$$

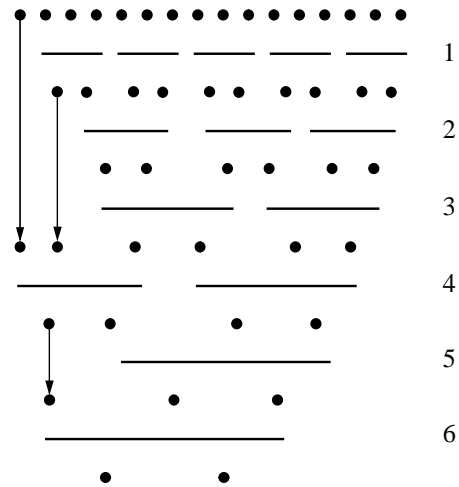
gate delays, including the initial AND gate delay to develop all bit products. The answer is thus 13, as computed directly in Part (a), for the  $4 \times 4$  case.

- 6.23. The number of reduction steps  $n$  to reduce  $k$  summands to 2 is given by  $k(2/3)^n = 2$ , because each step reduces 3 summands to 2. Then we have:

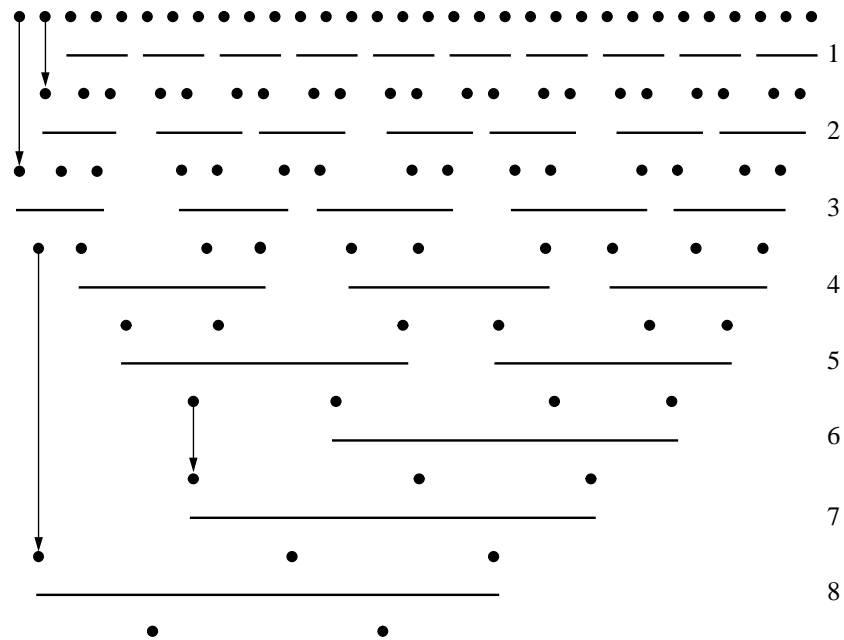
$$\begin{aligned} \log_2 k + n(\log_2 2 - \log_2 3) &= \log_2 2 \\ \log_2 k &= 1 + n(\log_2 3 - \log_2 2) \\ &= 1 + n(1.59 - 1) \\ n &= \frac{(\log_2 k) - 1}{0.59} \\ &= 1.7 \log_2 k - 1.7 \end{aligned}$$

This answer is only an approximation because the number of summands is not a multiple of 3 in each reduction step.

6.24. (a) Six CSA levels are needed:



(b) Eight CSA levels are needed:



(c) The approximation gives 5.1 and 6.8 CSA levels, compared to 6 and 8 from Parts (a) and (b).



6.25. (a)

+1.7	0	01111	101101
-0.012	1	01000	100010
+19	0	10011	001100
1/8	0	01100	000000

“Rounding” has been used as the truncation method in these answers.

(b) Other than exact 0 and  $\pm\infty$ , the smallest numbers are  $\pm 1.000000 \times 2^{-14}$  and the largest numbers are  $\pm 1.111111 \times 2^{15}$ .

(c) Assuming sign-and-magnitude format, the smallest and largest integers (other than 0) are  $\pm 1$  and  $\pm(2^{11} - 1)$ ; and the smallest and largest fractions (other than 0) are  $\pm 2^{-11}$  and approximately  $\pm 1$ .

(d)

$$\begin{aligned} A + B &= 0\ 10001\ 000000 \\ A - B &= 0\ 10001\ 110110 \\ A \times B &= 1\ 10010\ 001011 \\ A/B &= 1\ 10000\ 011011 \end{aligned}$$

“Rounding” has been used as the truncation method in these answers.

6.26. (a) Shift the mantissa of  $B$  right two positions, and tentatively set the exponent of the sum to 100001. Add mantissas:

$$\begin{array}{r} (A) \quad 1.1111111000 \\ (B) \quad 0.01001010101 \\ \hline 10.01001001101 \end{array}$$

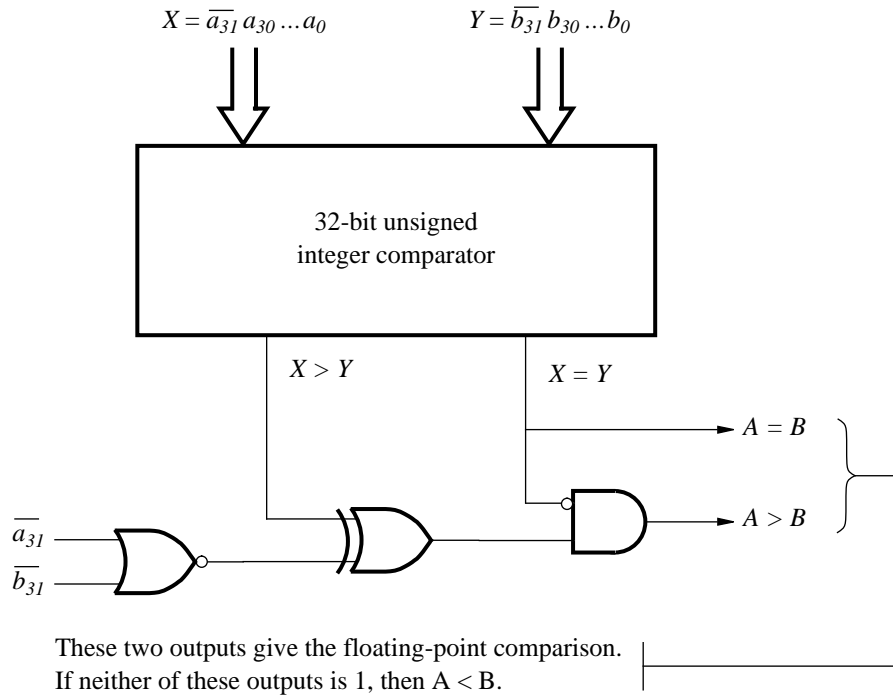
Shift right one position to put in normalized form: 1.001001001101 and increase exponent of sum to 100010. Truncate the mantissa to the right of the binary point to 9 bits by rounding to obtain 001001010. The answer is 0 100010 001001010.

(b)

$$\begin{aligned} \text{Largest} &\approx 2 \times 2^{31} \\ \text{Smallest} &\approx 1 \times 2^{-30} \end{aligned}$$

This assumes that the two end values, 63 and 0 in the excess-31 exponent, are used to represent infinity and exact 0, respectively.

- 6.27. Let  $A$  and  $B$  be two floating-point numbers. First, assume that  $S_A = S_B = 0$ . If  $E'_A > E'_B$ , considered as unsigned 8-bit numbers, then  $A > B$ . If  $E'_A = E'_B$ , then  $A > B$  if  $M_A > M_B$ . This means that  $A > B$  if the 31 bits after the sign in the representation for  $A$  is greater than the 31 bits representing  $B$ , when both are considered as integers. In the logic circuit shown below, all possibilities for the sum bit are also taken into account. In the circuit, let  $A = a_{31}a_{30} \dots a_0$  and  $B = b_{31}b_{30} \dots b_0$  be the two floating-point numbers to be compared.



- 6.28. Convert the given decimal mantissa into a binary floating-point number by using the integer facilities in the computer to implement the conversion algorithms in Appendix E. This will yield a floating-point number  $f_i$ . Then, using the computer's floating-point facilities, compute  $f_i \times t_i$ , as required.

- 6.29.  $(0.1)^{10} \Rightarrow (0.00011001100\dots)$

The signed, 8-bit approximations to this decimal number are:

Chopping:	$(0.1)_{10} = (0.0001100)_2$
Von Neumann Rounding:	$(0.1)_{10} = (0.0001101)_2$
Rounding:	$(0.1)_{10} = (0.0001101)_2$

- 6.30. Consider  $A - B$ , where  $A$  and  $B$  are 6-bit (normalized) mantissas of floating-point numbers. Because of differences in exponents,  $B$  must be shifted 6 positions before subtraction.

$$\begin{aligned} A &= 0.100000 \\ B &= 0.100001 \end{aligned}$$

After shifting, we have:

$$\begin{array}{rcl} A &= & 0.100000\ 000 \\ -B &= & \begin{array}{r} 0.000000\ 101 \\ \hline 0.011111\ 011 \end{array} \leftarrow \text{sticky bit} \\ \text{normalize} && 0.111110\ 110 \\ \text{round} && 0.111111 \leftarrow \text{correct answer (rounded)} \end{array}$$

With only 2 guard bits, we would have had:

$$\begin{array}{rcl} A &= & 0.100000\ 00 \\ -B &= & \begin{array}{r} 0.000000\ 11 \\ \hline 0.011111\ 01 \end{array} \\ \text{normalize} && 0.111110\ 10 \\ \text{round} && 0.111110 \end{array}$$

- 6.31. The binary versions of the decimal fractions  $-0.123$  and  $-0.1$  are not exact. Using 3 guard bits, with the last bit being the sticky bit, the fractions  $0.123$  and  $0.1$  are represented as:

$$\begin{aligned} 0.123 &= 0.00011\ 111 \\ 0.1 &= 0.00011\ 001 \end{aligned}$$

The three representations for both fractions using each of the three truncation methods are:

		Chop	Von Neumann	Round
-0.123:	Sign-and-magnitude	1.00011	1.00011	1.00100
	1's-complement	1.11100	1.11100	1.11011
	2's-complement	1.11101	1.11101	1.11100
-0.1:	Sign-and-magnitude	1.00011	1.00011	1.00011
	1's-complement	1.11100	1.11100	1.11100
	2's-complement	1.11101	1.11101	1.11101

6.32. The relevant truth table and logic equations are:

ADD(0) / SUBTRACT(1) (AS)	$S_A$	$S_B$	sign from 8-bit subtractor ( $8_s$ )	sign from 25-bit adder/ subtractor ( $25_s$ )	ADD/ SUB	$S_R$
0	0	0	0	0	0	0
			1	1		d
0	0	1	0	0	1	0
			1	1		d
0	1	0	0	0	1	1
			1	1		d
0	1	1	0	0	0	1
			1	1		d
1	0	0	0	0	1	1
			1	1		d
1	0	1	0	0	0	d
			1	1		d
1	1	0	0	0	0	1
			1	1		d
1	1	1	0	0	1	0
			1	1		d
these variables determine ADD/SUB			1	1		d

$S_A S_B$	00	01	11	10
ADD(0)/ SUBTRACT(1) (AS)	0	0	1	0
	1	1	0	1

$ADD/SUB = AS \oplus S_A \oplus S_B$

$S_B 8_s$	AS	$S_A$	00	01	11	10
00	0	0	1	1	0	0
01	0	0	1	1	0	0
11	1	1	0	0	1	1
10	0	1	1	1	0	0

$25_s = 0$

$S_B 8_s$	AS	$S_A$	00	01	11	10
00	d	0	d	1	d	1
01	d	d	d	d	d	d
11	d	d	d	d	d	d
10	1	d	0	d	d	d

$25_s = 1$

$$S_R = 25_s \bar{S}_A + \bar{25}_s S_A \bar{8}_s + AS \bar{S}_B 8_s + \bar{AS} S_B 8_s$$

6.33. The largest that  $n$  can be is 253 for normal values. The mantissas, including the leading bit of 1, are 24 bits long. Therefore, the output of the SHIFTER can be non-zero for  $n \leq 23$  only, ignoring guard bit considerations. Let  $n = n_7n_6 \dots n_0$ , and define an enable signal, EN, as  $EN = \bar{n}_7\bar{n}_6\bar{n}_5$ . This variable must be 1 for any output of the SHIFTER to be non-zero. Let  $m = m_{23}m_{22} \dots m_0$  and  $s_{23}s_{22} \dots s_0$  be the SHIFTER inputs and outputs, respectively. The largest network is required for output  $s_0$ , because any of the 24 input bits could be shifted into this output position. Define an intermediate binary vector  $i = i_{23}i_{22} \dots i_0$ . We will first shift  $m$  into  $i$  based on EN and  $n_4n_3$ . (Then we will shift  $i$  into  $s$ , based on  $n_2n_1n_0$ .) Only the part of  $i$  needed to generate  $s_0$  will be specified.

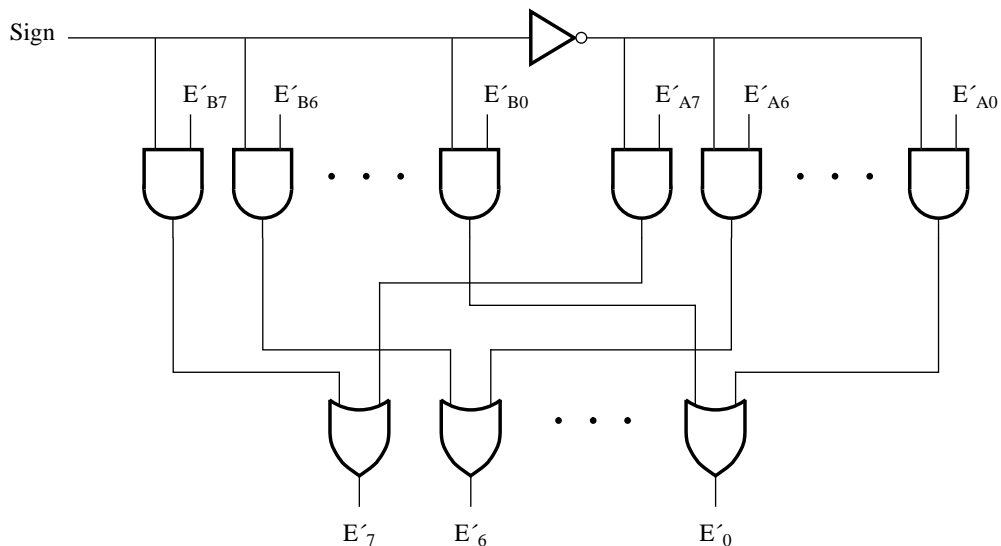
$$\begin{aligned}
i_7 &= ENn_4\bar{n}_3m_{23} + EN\bar{n}_4n_3m_{15} + EN\bar{n}_4\bar{n}_3m_7 \\
i_6 &= (\dots)m_{22} + (\dots)m_{14} + (\dots)m_6 \\
i_5 &= (\dots)m_{21} + (\dots)m_{13} + (\dots)m_5 \\
&\vdots \\
&\vdots \\
&\vdots \\
i_0 &= (\dots)m_{16} + (\dots)m_8 + (\dots)m_0
\end{aligned}$$

Gates with fan-in up to only 4 are needed to generate these 8 signals. Note that all bits of  $m$  are involved, as claimed. We now generate  $s_0$  from these signals and  $n_2n_1n_0$  as follows:

$$\begin{aligned}
s_0 &= n_2n_1n_0i_7 + n_2n_1\bar{n}_0i_6 + n_2\bar{n}_1n_0i_5 + n_2\bar{n}_1\bar{n}_0i_4 \\
&\quad + \bar{n}_2n_1n_0i_3 + \bar{n}_2n_1\bar{n}_0i_2 + \bar{n}_2\bar{n}_1n_0i_1 + \bar{n}_2\bar{n}_1\bar{n}_0i_0
\end{aligned}$$

Note that this requires a fan-in of 8 in the OR gate, so that 3 gates will be needed. Other  $s_i$  positions can be generated in a similar way.

6.34. (a)



(b) The SWAP network is a pair of multiplexers, each one similar to (a).

6.35. Let  $m = m_{24}m_{23} \dots m_0$  be the output of the adder/subtractor. The leftmost bit,  $m_{24}$ , is the overflow bit that could result from addition. (We ignore the handling of guard bits.) Derive a series of variables,  $z_i$ , as follows:

$$\begin{aligned}
 z_{-1} &= m_{24} \\
 z_0 &= \overline{m}_{24}m_{23} \\
 z_1 &= \overline{m}_{24}\overline{m}_{23}m_{22} \\
 &\vdots \\
 z_{23} &= \overline{m}_{24}\overline{m}_{23} \dots m_0 \\
 z_{24} &= \overline{m}_{24}\overline{m}_{23} \dots \overline{m}_0
 \end{aligned}$$

Note that exactly one of the  $z_i$  variables is equal to 1 for any particular  $m$  vector. Then encode these  $z_i$  variables, for  $-1 \leq i \leq 23$ , into a 6-bit signal representation for  $X$ , so that if  $z_i = 1$ , then  $X = i$ . The variable  $z_{24}$  signifies whether or not the resulting mantissa is zero.

- 6.36. Augment the 24-bit operand magnitudes entering the adder/subtractor by adding a sign bit position at the left end. Subtraction is then achieved by complementing the bottom operand and performing addition. Group corresponding bit-pairs from the two, signed, 25-bit operands into six groups of four bit-pairs each, plus one bit-pair at the left end, for purposes of deriving  $P_i$  and  $G_i$  functions. Label these functions  $P_6, G_6, \dots, P_0, G_0$ , from left-to-right, following the pattern developed in Section 6.2.

The lookahead logic must generate the group input carries  $c_0, c_4, c_8, \dots, c_{24}$ , accounting properly for the “end-around carry”. The key fact is that a carry  $c_i$  may have the value 1 because of a generate condition (i.e., some  $G_i = 1$ ) in a higher-order group as well as in a lower-order group. This observation leads to the following logic expressions for the carries:

$$\begin{aligned} c_0 &= G_6 + P_6 G_5 + \dots + P_6 P_5 P_4 P_3 P_2 P_1 G_0 \\ c_4 &= G_0 + P_0 G_6 + P_0 P_6 G_5 + \dots + P_0 P_6 P_5 P_4 P_3 P_2 G_1 \\ &\vdots \\ &\vdots \\ &\vdots \end{aligned}$$

Since the output of this adder is in 1’s-complement form, the sign bit determines whether or not to complement the remaining bits in order to send the magnitude  $M$  on to the “Normalize and Round” operation. Addition of positive numbers leading to overflow is a valid result, as discussed in Section 6.7.4, and must be distinguished from a negative result that may occur when subtraction is performed. Some logic at the left-end sign position solves this problem.